Recording Equalizer Amplifier for Stereo Cassette Decks

## Description

The CXA1598M/S is a bipolar IC developed for recording equalizer amplifier in analog cassette decks. Incorporating the filter circuit has eliminated the external inductor. Also, each of the six parameters required for equalizer amplifiers can be set independently with external resistance.

## Features

- Inductor (coil) is unnecessary
- The six parameters (low frequency gain, medium frequency gain, peaking gain, medium frequency compensation frequency, peaking frequency, and Q) required for recording equalizer amplifiers can be set independently with external resistance
- Low frequency boost is possible with an external capacitor
- Built-in recording mute function (requiring only an external time constant circuit to implement soft mute)
- Built-in 2 channels
- Small package


## Applications

Recording equalizer amplifier for stereo analog cassette decks


## Structure

Bipolar silicon monolithic IC

## Absolute Maximum Ratings

- Supply voltage Vcc

17
V

- Operating temperature
Topr $\quad-20$ to $+75 \quad{ }^{\circ} \mathrm{C}$
- Storage temperature

Tstg $\quad-65$ to $+150 \quad{ }^{\circ} \mathrm{C}$

- Allowable power dissipation

| Pd (CXA1598M) | 570 | mW |
| ---: | ---: | ---: |
| $(\mathrm{CXA} 1598 \mathrm{~S})$ | 880 | mW |

## Operating Conditions

Power supply
Dual power supplies (Vcc - Vee) $\pm 6.5$ to $8.0 \quad \mathrm{~V}$
Single power supply (Vcc)

## Block Diagram and Pin Configuration

CXA1598M


CXA1598S


Pin Description
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{Vcc}=7.0 \mathrm{~V}, \mathrm{VEE}=-7.0 \mathrm{~V}\right)$

| Pin No. |  | Symbol | Typical pinvoltage |  | I/O | Z (in) | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{array}{\|c\|} \hline \text { CXA } \\ 1598 \mathrm{M} \end{array}$ | $\begin{gathered} \text { CXA } \\ 15988 \end{gathered}$ |  | DC | AC |  |  |  |  |
| $\begin{aligned} & 1 \\ & 2 \\ & 3 \\ & 22 \\ & 23 \\ & 24 \end{aligned}$ | $\begin{array}{\|c} 1 \\ 2 \\ 3 \\ 20 \\ 21 \\ 22 \end{array}$ | $\begin{aligned} & \mathrm{fxQ} \\ & \mathrm{f} / \mathrm{Q} \\ & \mathrm{fM} \\ & \mathrm{GL} \\ & \mathrm{GH} \\ & \mathrm{GP} \end{aligned}$ | 1.2 V | - | 0 | - |  | Connection pins of resistors for setting the recording equalizer amplifier parameters. * Current input pins used to set the parameters for the recording equalizer amplifier. <br> * Setting currents for each parameter are generated by attaching resistors between these pins and the DGND pin. |
| 4 | 4 | FP CAL | 2.5 V | - | 1 | 54k |  | Peaking frequency calibration pin. <br> * Controlled with DC voltages of 0 to 5 V . High = Peaking frequency increased Low = Peaking frequency reduced <br> * Leave this pin open when not using the peaking frequency calibration function. |
| 19 | 17 | REC CAL | 2.5 V | - | 1 | 54k | (19) <br> (17) | Recording level calibration pin. <br> * Controlled with DC voltages of 0 to 5 V . <br> High = Recording level gain increased Low = Recording level gain reduced <br> * Leave this pin open when not using the recording level calibration function. |
| 5 | 5 | DGND | 0.0V | - | 1 | - |  | Connected to GND. |


| Pin No. |  | Symbol | Typical pinvoltage |  | I/O | Z (in) | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \text { CXXA } \\ 1598 \mathrm{M} \end{gathered}$ | $\begin{gathered} \text { CXA } \\ 15985 \end{gathered}$ |  | DC | AC |  |  |  |  |
| 6 | 6 | REC MUTE | 5.0 V | - | 1 | - |  | Recording mute <br> ON/OFF selection pin. <br> * Recording mute is controlled with DC voltages of 0 to 5 V . High = Recording mute OFF Low $=$ Recording mute ON <br> * Soft mute and fader can be switched over by changing the time constant of the external time constant circuit. |
| $\begin{array}{\|c} 8 \\ 17 \end{array}$ | $\begin{array}{\|c} 7 \\ 16 \end{array}$ | REC IN1 REC IN2 | 0.0 V | -18dBv | 1 | $50 \mathrm{k} \Omega$ |  | Recording equalizer amplifier input pin. |
| 9 | 8 | GND (VG) | 0.0V | - | 1 | - |  | Connect to GND for positive/negative dual power supplies. Vcc/2 (center potential) for a single power supply. (Connect a capacitor of $10 \mu \mathrm{~F}$ or more) |
| $\begin{array}{\|l} 10 \\ 15 \end{array}$ | $\begin{gathered} 9 \\ 14 \end{gathered}$ | BOOST1 BOOST2 | 0.0V | - | 1 | $9.5 \mathrm{k} \Omega$ |  | Connection pin of an external capacitor for low frequency boost. * When low frequency boost is unnecessary, connect to GND for positive/negative dual power supplies; connect a capacitor ( $3.3 \mu \mathrm{~F}$ or more) for a single power supply. |


| Pin No. |  | Symbol | Typical pin voltage |  | I/O | Z (in) | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{array}{\|c\|} \hline \text { CXA } \\ 1598 \mathrm{M} \end{array}$ | $\begin{aligned} & \text { CXA } \\ & 15985 \end{aligned}$ |  | DC | AC |  |  |  |  |
| 11 | 10 | Vee | -7.0V | - | 1 | - |  | Connect to the negative power supply for positive/negative dual power supplies. Connect to GND for a single power supply. |
| $\begin{array}{\|l} 12 \\ 13 \end{array}$ | $\begin{aligned} & 11 \\ & 12 \end{aligned}$ | REC OUT1 REC OUT2 | 0.0V | $-3.0 \mathrm{dBv}$ | 0 | - |  | Recording equalizer amplifier output pin. |
| 14 | 13 | Vcc | 7.0V | - | 1 | - |  | Positive power supply connection pin. |
| 16 | 15 | IREF | 1.2 V | - | 0 | - |  | Reference current setting pin for monolithic filter. <br> * The reference current can be set by attaching a resistor between this pin and DGND. |


| Pin No. |  | Symbol | Typical pin voltage |  | I/O | Z (in) | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{array}{c\|} \hline \text { CXA } \\ 1598 \mathrm{M} \end{array}$ | $\begin{gathered} \text { CXA } \\ 15985 \end{gathered}$ |  | DC | AC |  |  |  |  |
| 20 | 18 | GH CAL | 2.5 V | - | 1 | - |  | Medium frequency calibration pin. <br> * Controlled with DC voltages of 0 to 5 V . <br> High = Medium frequency level gain increased Low $=$ Medium frequency level gain reduced <br> * Leave this pin open when not using the medium frequency calibration function. |
| 21 | 19 | GP CAL | 2.5 V | - | 1 | - |  | High frequency calibration pin. <br> * Controlled with DC voltages of 0 to 5 V . <br> High = High frequency level gain increased Low = High frequency level gain reduced <br> * Leave this pin open when not using the high frequency calibration function. |

Electrical Characteristics
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{Vcc}=7.0 \mathrm{~V}, \mathrm{VEE}=-7.0 \mathrm{~V}\right)$

| Item | Conditions | Min. | Typ. | Max. | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Current consumption (IcC) | Standard settings <br> RGL: 36k/510k, RGH: 62k/220k, <br> RGP: 36k/110k, RfM: 39k/910k, <br> Rf/Q: 47k/750k, RfxQ: 47k/620k | 10.0 | 13.6 | 17.4 | mA |


|  | Item | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | GH-CAL characteristics 1 $(\mathrm{GH}-\mathrm{CAL}=5.0 \mathrm{~V})$ | $\mathrm{GH}-\mathrm{CAL}=5.0 \mathrm{~V}$ <br> RGH: 62k/220k, RGL•RGP: OPEN <br> RfM: 300k, Rf/Q: 18k, RfxQ: 12k <br> Input a 6.3 kHz signal ( -20 dB level down) and measure the amount of change compared to when the GH-CAL function is at the standard setting. | 4.7 | 5.7 | 6.7 | dB |
|  | GH-CAL characteristics 2 $(\mathrm{GH}-\mathrm{CAL}=0.0 \mathrm{~V})$ | $\mathrm{GH}-\mathrm{CAL}=0.0 \mathrm{~V}$ <br> RGH: 62k/220k, RGL•RGP: OPEN <br> RfM: 300k, Rf/Q: 18k, RfxQ: 12k <br> Input a 6.3 kHz signal ( -20 dB level down) and measure the amount of change compared to when the GH-CAL function is at the standard setting. | -5.5 | -4.5 | -3.5 | dB |
|  | GP-CAL characteristics 1 $(\mathrm{GP}-\mathrm{CAL}=5.0 \mathrm{~V})$ | $\mathrm{GP}-\mathrm{CAL}=5.0 \mathrm{~V}$ <br> RGP: 36k/110k, RGL•RGH: OPEN <br> RfM: 300k, Rf/Q: 47k/750k, RfxQ: 47k/620k <br> Input a signal ( -20 dB level down) and measure the amount of change compared to when the GP-CAL function is at the standard setting. | 3.9 | 5.4 | 6.9 | dB |
|  | GP-CAL characteristics 2 $(\mathrm{GP}-\mathrm{CAL}=0.0 \mathrm{~V})$ | $\mathrm{GP}-\mathrm{CAL}=0.0 \mathrm{~V}$ <br> RGP: 36k/110k, RGL•RGH: OPEN <br> RfM: 300k, Rf/Q: 47k/750k, RfxQ: 47k/620k <br> Input a signal (-20dB level down) and measure the amount of change compared to when the GP-CAL function is at the standard setting. | -5.8 | -4.3 | -2.8 | dB |
|  | FP-CAL characteristics 1 (FP-CAL = 5.0V) | $\text { FP-CAL }=5.0 \mathrm{~V}$ <br> Input a signal (-20dB level down) and measure the amount of change compared to when the FP-CAL function is at the standard setting. | 185 | 200 | 215 | \% |
|  | FP-CAL characteristics 2 $(F P-C A L=0.0 \mathrm{~V})$ | $\text { FP-CAL }=0.0 \mathrm{~V}$ <br> Input a signal ( -20 dB level down) and measure the amount of change compared to when the FP-CAL function is at the standard setting. | 36 | 46 | 56 | \% |
|  | fM medium frequency compensation frequency variable width |  | 0.3 | 2.4 | 10 | kHz |
|  | fp peaking frequency variable width |  | 10 | 17.8 | 50 | kHz |
|  | Peaking Q variable width |  | 2 | 4.2 | 7 |  |
|  | GL low frequency gain variable width |  | -5 | 0 | 8 | dB |
|  | GH medium frequency gain variable width |  | -10 | -3 | 11 | dB |
|  | GP peaking gain variable width |  | 10 | 20.5 | 30 | dB |
|  | fM medium frequency compensation frequency deviation | RGL: 36k/510k, RGH • RGP: OPEN or RGH: 62k/220k, RGL•RGP: OPEN <br> RfM: 39k/910k, Rf/Q: 18k, RfxQ: 12k | -15 | 0 | 15 | \% |
|  | fp peaking frequency deviation | RGP: 36k/110k, RGL•RGH: OPEN RfM: 300k, Rf/Q: 47k/750k, RfxQ: 47k/620k | -15 | 0 | 15 | \% |


| Item |  | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Peaking Q deviation | RGP: 36k//110k, RGL•RGH: OPEN <br> RfM: 300k, Rf/Q: 47k//750k, RfxQ: 47k//620k | -20 | 0 | 20 | \% |
|  | GL low frequency gain deviation | RGP: 36k//510k, RGH • RGP: OPEN RfM: 9.1k, Rf/Q: 18k, RfxQ: 12k | -0.5 | 0 | 0.5 | dB |
|  | GH medium frequency gain deviation | RGH: 62k//220k, RGL•RGP: OPEN RfM: 300k, Rf/Q: 18k, RfxQ: 12k | -0.8 | 0 | 0.8 | dB |
|  | GP peaking gain deviation | RGP: 36k//110k, RGL•RGH: OPEN <br> RfM: 300k, Rf/Q: 47k//750k, RfxQ: 47k/620k | -2.0 | 0 | 2.0 | dB |
|  | Input impedance | Pins 8 and 17 (CXA1598M) <br> Pins 7 and 16 (CXA1598S) | 40 | 50 | 60 | k $\Omega$ |

Note: Unless otherwise specified, RGL, RGH, RGP, RfM, Rf/Q, and RfxQ settings are the characteristics when set to the standard settings.
Electrical Characteristics Measurement Circuit (CXA1598S)


Application Circuit for Positive/Negative Dual Power Supplies (CXA1598S)

Application Circuit for Positive/Negative Dual Power Supplies (CXA1598M)

Application Circuit for a Single Power Supply (CXA1598S)

Application Circuit for a Single Power Supply (CXA1598M)


## Description of Operation

## 1. Recording equalizer amplifier

The primary features of the CXA1598 recording equalizer amplifier are that by taking full advantage of monolithic filter technology, an LC resonance circuit consisting of a coil and capacitor normally required for high frequency compensation is dispensed with and medium and low frequency sensitivity compensation is performed with its internal filter alone. In addition, the six parameters (low frequency gain, medium frequency gain, peaking gain, medium frequency compensation frequency, peaking frequency, and $Q$ ) required for recording equalizer amplifiers can be set as desired simply by attaching resistors to the GL, GH, GP, fM, f/Q, and fxQ pins.

This IC has the circuit configuration shown in Fig. 1 to provide the optimum frequency response required for recording equalizer amplifiers.


Fig. 1. CXA1598M/S functional circuit block diagram

## 2. Low frequency boost

The CXA1598 implements low frequency boost simply by attaching an external capacitor to the BOOST pins. Signals are boosted by approximately 6dB. The boost cut-off frequency can be freely set with the value of the external capacitor.


Fig. 2. CXA1598M/S low frequency boost frequency response

## 3. Recording mute function

The CXA1598 contains a built-in recording mute circuit which varies the recording equalizer amplifier gain according to the magnitude of the DC voltage applied to the REC MUTE pin just like an electronic volume control. Also, any desired soft mute or fader can be freely set depending on momentary changes in the DC voltage applied to the REC MUTE pin. Fig. 3 illustrates the recording mute waveforms.


Fig. 3. Recording mute waveform

## 4. Recording level calibration function

The CXA1598 allows the recording level to be finely adjusted with a DC voltage. The recording equalizer amplifier gain can be varied by approximately $\pm 5 \mathrm{~dB}$ simply by applying a DC voltage to the REC CAL pin. When not using the recording level calibration function, simply leave the REC CAL pin open, and the REC CAL pin is matched to the internal reference voltage $(2.5 \mathrm{~V})$, with the recording level set for the standard output gain.

## 5. Medium frequency equalizer amplifier calibration function

The CXA1598 allows the medium frequency equalizer amplifier characteristics to be finely adjusted with a DC voltage. By simply applying a DC voltage to the GH CAL pin, the medium frequency equalizer amplifier gain can be varied by approximately $\pm 4 \mathrm{~dB}$. When not using this calibration function, simply leave the GH CAL pin open, and the GH CAL pin is matched to the internal reference voltage ( 2.5 V ), with the medium frequency equalizer amplifier characteristics set for the standard output gain.

## 6. High frequency equalizer amplifier calibration function

The CXA1598 allows the high frequency equalizer amplifier characteristics to be finely adjusted with a DC voltage. By simply applying a DC voltage to the GP CAL pin, the high frequency equalizer amplifier gain can be varied by approximately $\pm 4 \mathrm{~dB}$. Also, when not using this calibration function, simply leave the GP CAL pin open, and the GP CAL pin is matched to the internal reference voltage $(2.5 \mathrm{~V})$, with the high frequency equalizer amplifier characteristics set for the standard output gain.

## 7. fp peaking frequency calibration function

The CXA1598 allows the fp peaking frequency to be finely adjusted with a DC voltage. By simply applying a DC voltage to the FP CAL pin, the fp peaking frequency can be varied by approximately $46 \%$ to $200 \%$. Also, when not using this calibration function, simply leave the FP CAL pin open, and the FP CAL pin is matched to the internal reference voltage ( 2.5 V ), with the fp peaking frequency response set for the standard fp peaking frequency.


Fig. 4. Conceptual diagram of recording level/medium frequency equalizer amplifier/high frequency equalizer amplifier/fp peaking frequency calibration functions

## Control Voltage for Each Control Pin

| $\begin{aligned} & \text { Pin } \\ & \text { NO. } \end{aligned}$ | Pin Name | Pin voltage [V], referenced to DGND |  |  |  |  | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 0.0 | 0.5 | 2.5 | 4.5 | 5.0 |  |
| $\begin{gathered} 4 \\ (4) \end{gathered}$ | FP CAL | Reduce $\lll \lll<$ Increase |  |  |  |  | Amount of $\mathfrak{f p}$ peaking frequency change [\%] compared to when FP CAL is at the standard setting. |
|  |  | 46 | - | - | - | 200 |  |
| $\begin{gathered} 6 \\ (6) \end{gathered}$ | REC MUTE | Reduce $\lll \lll<$ Increase |  |  |  |  | REC OUT attenuation [dB] compared to when REC MUTE is at the standard setting. $\mathrm{f}=1 \mathrm{kHz}$ |
|  |  | - | -100 | -4.5 | - | - |  |
| $\begin{gathered} 17 \\ (19) \end{gathered}$ | REC CAL | Reduce \lll \lll < lncrease |  |  |  |  | Amount of change [dB] compared to when REC CAL is at the standard setting.$\mathrm{f}=315 \mathrm{~Hz}$ |
|  |  | -6.5 | - | - | - | 6.0 |  |
| $\begin{gathered} 18 \\ (20) \end{gathered}$ | GH CAL | Reduce $\lll \lll<$ Increase |  |  |  |  | Amount of GH medium frequency gain change [dB] compared to RGH standard. RGL, RGP: OPEN |
|  |  | -4.5 | - | - | - | 5.7 |  |
| $\begin{gathered} 19 \\ (21) \end{gathered}$ | GP CAL | Reduce $\lll \lll<$ Increase |  |  |  |  | Amount of GP peaking frequency gain change [dB] compared to RGP standard. RGL, RGH: OPEN |
|  |  | -4.2 | - | - | - | 5.4 |  |

## 8. Mode control methods

Refer to the application circuits shown in Figs. 5 and 6 for mode control methods using a manual switch.
When tape mode is implemented with logic, use the same ground for the $27 \mathrm{k} \Omega$ resistance connected to the common pin (analog switch connection) of the used analog switch IC and to the DGND and IREF pins.
Figs. 5 and 6 show examples when using the 4051B (8-channel multiplexer/demultiplexer).


Fig. 5. For positive/negative dual power supplies


Fig. 6. For a single power supply

## 9. Temperature characteristics and accuracy of the recording equalizer amplifier

The temperature and cut-off frequency of the CXA1598 depend on the external resistance connected to the IREF, GL, GH, GP, fM, f/Q, and fxQ pins. For low frequency boost, however, the cut-off frequency becomes uneven depending on the temperature characteristics or unevenness of the internal resistance since its time constant is configured by the product of an external capacitor and the internal resistance.
Also, the recording equalizer amplifier frequency response depends on unevenness in the absolute, as well as relative values of the internal capacitance. Furthermore, the high frequency response indicates a high element sensitivity at the filter because the band-pass filter $Q$ is high. Compared to low frequency, although the unevenness inherent in the IC is more likely to occur, this occurs relatively, and not individually for channels 1 and 2 .

## Notes on Operation

## 1. Power supply

The CXA1598 is designed basically for positive/negative dual power supplies, and can also operate with a single power supply. Connect the power supplies for each case as shown below:

|  | Vcc 14pin (13pin) | VEE 11pin (10pin) | GND 9pin (8pin) | DGND 5pin (5pin) |
| :--- | :---: | :---: | :---: | :---: |
| Positive/negative <br> dual power supplies | Positive power supply | Negative power supply | GND | GND |
| Single power supply | Power supply | GND | —_* $^{*}$ | GND |

Pin Nos. in parentheses are those for the CXA1598S.

* For a single power supply, connect a decoupling capacitor ( $10 \mu \mathrm{~F}$ or more) to the GND (VG) pin. The ripple rejection ratio depends on the capacitance of this capacitor.


## 2. Low frequency boost

The CXA1598 can implement low frequency boost simply by connecting a capacitor to the BOOST pins. Although the boost is fixed to 6 dB , the time constant which determines the cut-off frequency can be set to any desired value depending on the external capacitor. The pole ( $\mathrm{f}_{1}$ ) and zero ( $\mathrm{f}_{2}$ ) shown in Fig. 3. Low frequency boost frequency response can be expressed, with the external capacitor assumed to be Св, as follows:

$$
\begin{aligned}
\mathrm{f}_{1} & =\frac{\mathrm{R} 13+\mathrm{R} 14}{2 \pi \cdot \mathrm{CB}_{B} \cdot(\mathrm{R} 13 \cdot \mathrm{R} 14+\mathrm{R} 14 \cdot \mathrm{R} 15+\mathrm{R} 15 \cdot \mathrm{R} 13)} \\
& =\frac{1}{2 \pi \cdot \mathrm{CB}_{B} \cdot(\mathrm{R} 13 \cdot \mathrm{R} 14 /(\mathrm{R} 13+\mathrm{R} 14)+\mathrm{R} 15)}=\frac{1}{2 \pi \cdot \mathrm{CB}_{B} \cdot(9.53 \mathrm{k} \Omega)}[\mathrm{Hz}] \\
& =\frac{1}{2 \pi \cdot \mathrm{CB}_{B} \cdot \mathrm{R} 15}=\frac{1}{2 \pi \cdot \mathrm{CB}_{\mathrm{B}} \cdot(4.8 \mathrm{k} \Omega)}[\mathrm{Hz}]
\end{aligned}
$$

When not using low frequency boost, follow the procedure described below.
© For positive/negative dual power supplies
Connect the BOOST pins to GND.
© For single power supply
Connect a fairly large capacitor ( $3.3 \mu \mathrm{~F}$ or more) to the BOOST pins or simply leave the BOOST pins open. If the BOOST pins are left open, note that the output level increases by 6 dB , so the input level should be set 6 dB down. The CXA1598 is basically designed for positive/negative dual power supplies and the BOOST pins cannot be easily connected to GND as in the case of positive/negative dual power supplies.

## 3. Resistance connected to the IREF pin as well as the GL, GH, GP, fM, $\mathrm{f} / \mathrm{Q}$, and fxQ pins

The recording equalizer amplifier frequency response is determined by the resistance connected to the IREF pin as well as the GL, GH, GP, fM, $f / Q$, and $f x Q$ pins. This means that the accuracy of the recording equalizer amplifier frequency response is determined by the resistance connected to these pins. Therefore, the resistors used for this purpose must be free of unevenness and have excellent temperature characteristics (e.g., a metallic film resistor).

Example of Representative Characteristics



RGP gain characteristics




Output level vs. Mute characteristics 1


Output level vs. Mute characteristics 2


Current consumption vs. Supply voltage


Output level vs. Mute characteristics 3



Output level vs. REC CAL voltage


Output level vs. GH CAL voltage


Output level vs. GP CAL voltage

fp peaking frequency vs. FP CAL pin voltage


REC CAL and GH CAL frequency response


REC CAL and GP CAL frequency response


REC CAL and FP CAL frequency response


## CXA1598M



| SONY CODE | SOP-24P-L01 |
| :--- | :--- |
| EIAJ CODE | *SOP024-P-0300-A |
| JEDEC CODE | - |

PACKAGE STRUCTURE

| MOLDING COMPOUND | EPOXY/PHENOL RESIN |
| :--- | :--- |
| LEAD TREATMENT | SOLDER PLATING |
| LEAD MATERIAL | COPPER ALLOY/ 42ALLOY |
| PACKAGE WEIGHT | 0.3 g |

CXA1598S
22PIN SDIP (PLASTIC)


